

Phase Error of Kim's Differential Atmospheric Phase Monitor
if the SNR is Unity or Greater
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The (real) signal voltage from the Satellite which is picked up by either antenna and band limited within the receiving system can be represented by the narrow band form $S(t) = s(t) \cdot \cos(\omega t - \phi(t))$ where $s(t)$ and $\phi(t)$ represent the amplitude and phase and vary at a rate of the inverse bandwidth. Similarly the noise voltage from an antenna can be represented as $N(t) = n_1(t) \cdot \cos(\omega t - \phi_1(t))$. If we assume that the delays have been set so that the satellite is at the center of the white light fringe, the output of I from the I/Q demodulator will be:

$$\begin{aligned} I &\propto \int_0^T [S(t) + N_1(t)] \cdot [S(t) + N_2(t)] dt \\ &= \int_0^T [S^2(t) + S(t) \cdot (N_1(t) + N_2(t)) + N_1(t) \cdot N_2(t)] dt \\ &= \int_0^T [s^2(t) \cdot \cos^2(\omega t - \phi(t)) \\ &\quad + s(t) \cdot \cos(\omega t - \phi(t)) \cdot (n_1(t) \cdot \cos(\omega t - \phi_1(t)) + n_2(t) \cdot \cos(\omega t - \phi_2(t))) \\ &\quad + n_1(t) \cdot \cos(\omega t - \phi_1(t)) + n_2(t) \cdot \cos(\omega t - \phi_2(t))] dt \end{aligned}$$

Only the $S^2(t)$ term integrates coherently, and the noise from the other terms will be small and may be ignored, so:

$$I \propto s^2$$

where s now refers to the rms of $s(t)$.

The Q output will be similar:

$$\begin{aligned} Q &\propto \int_0^T [s^2(t) \cdot \cos(\omega t - \phi(t)) \cdot \sin(\omega t - \phi(t)) \\ &\quad + s(t) \cdot \cos(\omega t - \phi(t)) \cdot (n_1(t) \cdot \cos(\omega t - \phi_1(t)) + n_2(t) \cdot \sin(\omega t - \phi_2(t))) \\ &\quad + n_1(t) \cdot \cos(\omega t - \phi_1(t)) + n_2(t) \cdot \sin(\omega t - \phi_2(t))] dt \end{aligned}$$

except that the $S^2(t)$ term will integrate to zero since one of the cosines has been converted to a sine. The remaining terms will integrate to approximately zero and we can apply the usual radiometer formula to determine the rms deviation of Q from zero:

$$Q_{rms} \propto \sqrt{\frac{2s^2 \cdot n^2 + n^4}{B\tau}}$$

where s and n refer to rms values and the two noise powers have been assumed to be equal. B is the IF bandwidth. The noise in fringe phase angle measurement (radians) is Q_{rms} / I :

$$\Phi_{rms} = Q_{rms} / I = \sqrt{\frac{2n^2/s^2 + n^4/s^4}{B\tau}}$$

In the usual radio astronomy case, the noise is much larger than the signal and this formula reduces to:

$$\Phi_{rms} = \frac{n^2}{s^2 \cdot \sqrt{B\tau}}$$

which is the usual ratio of noise power to signal power divided by $\sqrt{B\tau}$.

Kim calculates a signal to noise ratio of 5 dB ($n^2/s^2 = .3126$) for one satellite using a 1.6m dish. He gives $\sqrt{B\tau}$ as 4582 when sampling the phase at 20 Hz. Using the formula above and those conditions, $\Phi_{rms} = 0.0107$ deg.